



Observable state space realizations for multivariable systems[☆]

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ABSTRACT

This paper derives two canonical state space forms (i.e., the observer canonical form and the observability canonical form) from multiple-input multiple-output systems described by difference equations. The state space model is expressed by the first-order difference equation and is equivalent to the input–output representation. More specifically, by setting the different state variables, the difference equations or the input–output representations can be transformed into two observable canonical forms and the canonical state space model can be also transformed into the difference equations. Finally, two examples are given.

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1. Introduction

State space models play a significant role in system modeling and identification, adaptive control, and system analysis [1–5]. The state space model can describe a physical dynamic system with a set of the first-order differential equations or difference equations which involve the system input, output and state variables. The states represent the internal behavior of a system and the state space model maps the relationship from inputs to outputs. For multiple-input multiple-output systems (i.e., multivariable systems), the input and output are in a form of vectors. The state space representation provides a convenient and compact way to model and analyze systems. The process of transforming the differential equations (difference equations) or transfer function matrices to state space models is called the realization. The state-space realizations contain four basic canonical forms: (1) the observer canonical form and (2) the observability canonical form, which are observable (i.e., each state variable of the internal system can be completely determined by the output); (3) the controller canonical form and (4) the controllability canonical form, which are controllable (i.e., each state variable of the internal system can be completely determined by the input).

The difference equations are important because they act as mathematical models describing real life systems in computer science, economics, ecology [6–12]. In this literature, Kurbanli et al. studied the behavior of the positive solutions of the system of rational difference equations [13]; Iričanin discussed the global stability of some classes of higher-order nonlinear difference equations [14]. In the area of state space model identification, Schön et al. presented a system identification algorithm for the nonlinear state-space models [15]; Rachid et al. discussed the multivariable fractional system approximation with initial conditions using the integral state space representation [16]; Ding et al. presented the hierarchical state space model identification method for general dual-rate sampled-data systems [17] and for non-uniformly sampled-data systems [18–20].

Multivariable systems have received much attention in system analysis and control [21,22], system modeling and identification [23–32], and signal processing and parameter estimation [33–37]. Many methods discussed parameter

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identification problems of the input–output representations from state space models [38–48], including the multi-innovation identification algorithms [49–56], the maximum likelihood least squares algorithms [57–59], the iterative estimation algorithms [60–67] and the auxiliary model based identification algorithms [68–70]. This paper discusses the transformation relationships between the difference equations and the observer or observability canonical state space models for multivariable systems [17,18]. The proposed models could be applied to multivariable networked control system design and filtering [71–73].

Briefly, this paper is organized as follows. Sections 2 and 3 discuss the transformation between the difference equation and the observer canonical form. Sections 4 and 5 describe the transformation between the difference equation and the observability canonical form. Section 6 provides two examples to illustrate the proposed methods. Finally, we offer several concluding remarks in Section 7.

2. The observer canonical form from the difference equation

Systems with more than one input and more than one output are known as multiple-input multiple-output systems and the systems can be described by the difference equations. This section derives the state space model from a difference equation or an input–output representation.

Consider a linear dynamical system described by difference equation:

$$\begin{aligned} \mathbf{y}(t) + \mathbf{A}_1\mathbf{y}(t-1) + \mathbf{A}_2\mathbf{y}(t-2) + \cdots + \mathbf{A}_n\mathbf{y}(t-n) \\ = \mathbf{B}_1\mathbf{u}(t-1) + \mathbf{B}_2\mathbf{u}(t-2) + \mathbf{B}_3\mathbf{u}(t-3) + \cdots + \mathbf{B}_n\mathbf{u}(t-n), \end{aligned} \quad (1)$$

where $\mathbf{u}(t) \in \mathbb{R}^r$ and $\mathbf{y}(t) \in \mathbb{R}^m$ denote the input and output vectors of the system, t represents the time variable, $\mathbf{A}_i \in \mathbb{R}^{m \times m}$ and $\mathbf{B}_i \in \mathbb{R}^{m \times r}$ are coefficient matrices in z^{-1} .

Let \mathbf{I} be an identity matrix of appropriate sizes and z be a unit forward shift operator: $z\mathbf{x}(t) = \mathbf{x}(t+1)$ or $z^{-1}\mathbf{x}(t) = \mathbf{x}(t-1)$, $\mathbf{A}(z)$ and $\mathbf{B}(z)$ are the matrix polynomials and are defined as

$$\begin{aligned} \mathbf{A}(z) &:= \mathbf{I} + \mathbf{A}_1z^{-1} + \mathbf{A}_2z^{-2} + \cdots + \mathbf{A}_nz^{-n}, \\ \mathbf{B}(z) &:= \mathbf{B}_1z^{-1} + \mathbf{B}_2z^{-2} + \mathbf{B}_3z^{-3} + \cdots + \mathbf{B}_nz^{-n}. \end{aligned}$$

Using the properties of the shift operator, Eq. (1) can be rewritten as

$$(\mathbf{I} + \mathbf{A}_1z^{-1} + \mathbf{A}_2z^{-2} + \cdots + \mathbf{A}_nz^{-n})\mathbf{y}(t) = (\mathbf{B}_1z^{-1} + \mathbf{B}_2z^{-2} + \mathbf{B}_3z^{-3} + \cdots + \mathbf{B}_nz^{-n})\mathbf{u}(t).$$

That is

$$\mathbf{A}(z)\mathbf{y}(t) = \mathbf{B}(z)\mathbf{u}(t),$$

or

$$\mathbf{y}(t) = \mathbf{A}^{-1}(z)\mathbf{B}(z)\mathbf{u}(t), \quad (2)$$

where transfer matrix $\mathbf{G}(z) := \mathbf{A}^{-1}(z)\mathbf{B}(z) \in \mathbb{R}^{m \times r}$ from the input $\mathbf{u}(t)$ to the output $\mathbf{y}(t)$ is a strictly proper rational fraction matrix.

The high-order difference equations are not convenience in applications and is often transformed into a set of the first-order difference equations. The details are as follows.

Define the state variables,

$$\begin{aligned} \mathbf{x}_1(t) &:= \mathbf{y}(t), \\ \mathbf{x}_2(t) &:= \mathbf{y}(t+1) + \mathbf{A}_1\mathbf{y}(t) - \mathbf{B}_1\mathbf{u}(t) \\ &= \mathbf{x}_1(t+1) + \mathbf{A}_1\mathbf{x}_1(t) - \mathbf{B}_1\mathbf{u}(t), \end{aligned} \quad (3)$$

$$\begin{aligned} \mathbf{x}_3(t) &:= \mathbf{y}(t+2) + \mathbf{A}_1\mathbf{y}(t+1) + \mathbf{A}_2\mathbf{y}(t) - \mathbf{B}_1\mathbf{u}(t+1) - \mathbf{B}_2\mathbf{u}(t) \\ &= \mathbf{x}_2(t+1) + \mathbf{A}_2\mathbf{x}_1(t) - \mathbf{B}_2\mathbf{u}(t), \end{aligned} \quad (4)$$

\vdots

$$\begin{aligned} \mathbf{x}_n(t) &:= \mathbf{y}(t+n-1) + \mathbf{A}_1\mathbf{y}(t+n-2) + \mathbf{A}_2\mathbf{y}(t+n-3) + \cdots + \mathbf{A}_{n-1}\mathbf{y}(t) \\ &\quad - \mathbf{B}_1\mathbf{u}(t+n-2) - \mathbf{B}_2\mathbf{u}(t+n-3) - \mathbf{B}_3\mathbf{u}(t+n-4) - \cdots - \mathbf{B}_{n-1}\mathbf{u}(t) \end{aligned} \quad (5)$$

$$= \mathbf{x}_{n-1}(t+1) + \mathbf{A}_{n-1}\mathbf{x}_1(t) - \mathbf{B}_{n-1}\mathbf{u}(t). \quad (6)$$

From Eqs. (3), (4) and (6), we have

$$\begin{cases} \mathbf{x}_1(t+1) = -\mathbf{A}_1\mathbf{x}_1(t) + \mathbf{x}_2(t) + \mathbf{B}_1\mathbf{u}(t), \\ \mathbf{x}_2(t+1) = -\mathbf{A}_2\mathbf{x}_1(t) + \mathbf{x}_3(t) + \mathbf{B}_2\mathbf{u}(t), \\ \vdots \\ \mathbf{x}_{n-1}(t+1) = -\mathbf{A}_{n-1}\mathbf{x}_1(t) + \mathbf{x}_n(t) + \mathbf{B}_{n-1}\mathbf{u}(t). \end{cases} \quad (7)$$

Replacing t in (1) with $t + n$ gives

$$\begin{aligned} & \mathbf{y}(t+n) + \mathbf{A}_1 \mathbf{y}(t+n-1) + \mathbf{A}_2 \mathbf{y}(t+n-2) + \cdots + \mathbf{A}_{n-1} \mathbf{y}(t+1) \\ & - \mathbf{B}_1 \mathbf{u}(t+n-1) - \mathbf{B}_2 \mathbf{u}(t+n-2) - \mathbf{B}_3 \mathbf{u}(t+n-3) - \cdots - \mathbf{B}_{n-1} \mathbf{u}(t+1) \\ & = -\mathbf{A}_n \mathbf{y}(t) + \mathbf{B}_n \mathbf{u}(t). \end{aligned}$$

Replacing t in (6) with $t + 1$ leads to the following equation,

$$\begin{aligned} \mathbf{x}_n(t+1) &= \mathbf{y}(t+n) + \mathbf{A}_1 \mathbf{y}(t+n-1) + \mathbf{A}_2 \mathbf{y}(t+n-2) + \cdots + \mathbf{A}_{n-1} \mathbf{y}(t+1) \\ & - \mathbf{B}_1 \mathbf{u}(t+n-1) - \mathbf{B}_2 \mathbf{u}(t+n-2) - \mathbf{B}_3 \mathbf{u}(t+n-3) - \cdots - \mathbf{B}_{n-1} \mathbf{u}(t+1) \\ & = -\mathbf{A}_n \mathbf{y}(t) + \mathbf{B}_n \mathbf{u}(t) \\ & = -\mathbf{A}_n \mathbf{x}_1(t) + \mathbf{B}_n \mathbf{u}(t). \end{aligned} \quad (8)$$

Combining (7) and (8) gives the following state space model,

$$\begin{cases} \begin{bmatrix} \mathbf{x}_1(t+1) \\ \mathbf{x}_2(t+1) \\ \vdots \\ \mathbf{x}_n(t+1) \end{bmatrix} = \begin{bmatrix} -\mathbf{A}_1 & \mathbf{I} & \mathbf{0} & \cdots & \mathbf{0} \\ -\mathbf{A}_2 & \mathbf{0} & \mathbf{I} & \ddots & \vdots \\ \vdots & \vdots & \vdots & \ddots & \mathbf{0} \\ -\mathbf{A}_{n-1} & \mathbf{0} & \cdots & \mathbf{0} & \mathbf{I} \\ -\mathbf{A}_n & \mathbf{0} & \cdots & \mathbf{0} & \mathbf{0} \end{bmatrix} \begin{bmatrix} \mathbf{x}_1(t) \\ \mathbf{x}_2(t) \\ \vdots \\ \mathbf{x}_n(t) \end{bmatrix} + \begin{bmatrix} \mathbf{B}_1 \\ \mathbf{B}_2 \\ \vdots \\ \mathbf{B}_n \end{bmatrix} \mathbf{u}(t), \\ \mathbf{y}(t) = \mathbf{x}_1(t) = [\mathbf{I}, \mathbf{0}, \mathbf{0}, \dots, \mathbf{0}] \begin{bmatrix} \mathbf{x}_1(t) \\ \mathbf{x}_2(t) \\ \vdots \\ \mathbf{x}_n(t) \end{bmatrix}. \end{cases} \quad (9)$$

Or

$$\begin{cases} \mathbf{x}_0(t+1) = \mathbf{A}_0 \mathbf{x}_0(t) + \mathbf{B}_0 \mathbf{u}(t), \\ \mathbf{y}(t) = \mathbf{C}_0 \mathbf{x}_0(t), \end{cases} \quad (10)$$

$$\begin{aligned} \mathbf{A}_0 &:= \begin{bmatrix} -\mathbf{A}_1 & \mathbf{I} & \mathbf{0} & \cdots & \mathbf{0} \\ -\mathbf{A}_2 & \mathbf{0} & \mathbf{I} & \ddots & \vdots \\ \vdots & \vdots & \vdots & \ddots & \mathbf{0} \\ -\mathbf{A}_{n-1} & \mathbf{0} & \cdots & \mathbf{0} & \mathbf{I} \\ -\mathbf{A}_n & \mathbf{0} & \cdots & \mathbf{0} & \mathbf{0} \end{bmatrix} \in \mathbb{R}^{(mn) \times (mn)}, & \mathbf{B}_0 &:= \begin{bmatrix} \mathbf{B}_1 \\ \mathbf{B}_2 \\ \vdots \\ \mathbf{B}_n \end{bmatrix} \in \mathbb{R}^{(mn) \times r}, \\ \mathbf{C}_0 &:= [\mathbf{I}, \mathbf{0}, \mathbf{0}, \dots, \mathbf{0}] \in \mathbb{R}^{m \times (mn)}, & \mathbf{A}_i &\in \mathbb{R}^{m \times m}, & \mathbf{B}_i &\in \mathbb{R}^{m \times r}, \end{aligned}$$

where $\mathbf{x}_0(t) := [\mathbf{x}_1(t), \mathbf{x}_2(t), \dots, \mathbf{x}_n(t)]^T \in \mathbb{R}^{mn}$ is the state vector, \mathbf{A}_0 is the left-column block companion matrix [74,75], \mathbf{B}_0 is an arbitrary matrix and \mathbf{C}_0 is a matrix whose the first $m \times m$ block is an identity matrix and the rest is a zero matrix.

This state space model with \mathbf{A}_0 and \mathbf{C}_0 special structures is called the observer canonical form. The parameters in the observer canonical form are corresponding to that in the difference equation in (1) or the transfer function in (2).

3. The observer canonical form to the difference equation

This section is to transform the observer canonical form in (9) to its equivalent input–output representation. The basic idea is to eliminate the state vector $\mathbf{x}(t)$ in (9) and to obtain an input–output representation or difference equations.

Let

$$\mathbf{Q} := z\mathbf{I} - \mathbf{A}_0 = \begin{bmatrix} z\mathbf{I} + \mathbf{A}_1 & -\mathbf{I} & \mathbf{0} & \cdots & \mathbf{0} \\ \mathbf{A}_2 & z\mathbf{I} & -\mathbf{I} & \ddots & \vdots \\ \vdots & \mathbf{0} & \ddots & \ddots & \mathbf{0} \\ \mathbf{A}_{n-1} & \vdots & \ddots & z\mathbf{I} & -\mathbf{I} \\ \mathbf{A}_n & \mathbf{0} & \cdots & \mathbf{0} & z\mathbf{I} \end{bmatrix}, \quad \mathbf{Q}^{-1} := \begin{bmatrix} \mathbf{q}_{11} & \mathbf{q}_{12} & \mathbf{q}_{13} & \cdots & \mathbf{q}_{1n} \\ * & * & * & \cdots & * \\ \vdots & \vdots & \vdots & & \vdots \\ * & * & * & \cdots & * \end{bmatrix}.$$

According to $\mathbf{Q}^{-1}\mathbf{Q} = \mathbf{I}$, we have

$$\begin{bmatrix} q_{11} & q_{12} & q_{13} & \cdots & q_{1n} \\ * & * & * & \cdots & * \\ \vdots & \vdots & \vdots & & \vdots \\ * & * & * & \cdots & * \end{bmatrix} \begin{bmatrix} z\mathbf{I} + \mathbf{A}_1 & -\mathbf{I} & \mathbf{0} & \cdots & \mathbf{0} \\ \mathbf{A}_2 & z\mathbf{I} & -\mathbf{I} & \ddots & \vdots \\ \vdots & \mathbf{0} & \ddots & \ddots & \mathbf{0} \\ \mathbf{A}_{n-1} & \vdots & \ddots & z\mathbf{I} & -\mathbf{I} \\ \mathbf{A}_n & \mathbf{0} & \cdots & \mathbf{0} & z\mathbf{I} \end{bmatrix} = \begin{bmatrix} \mathbf{I} & \mathbf{0} & \cdots & \mathbf{0} \\ \mathbf{0} & \mathbf{I} & \cdots & \mathbf{0} \\ \vdots & \vdots & \ddots & \vdots \\ \mathbf{0} & \mathbf{0} & \cdots & \mathbf{I} \end{bmatrix}.$$

Expanding this matrix equation, its first row gives the following equations:

$$\begin{cases} q_{11}(z\mathbf{I} + \mathbf{A}_1) + q_{12}\mathbf{A}_2 + q_{13}\mathbf{A}_3 + \cdots + q_{1n}\mathbf{A}_n = \mathbf{I}, \\ -q_{11} + q_{12}z = \mathbf{0}, \\ -q_{12} + q_{13}z = \mathbf{0}, \\ \vdots \\ -q_{1n-1} + q_{1n}z = \mathbf{0}, \end{cases}$$

which can be equivalently rewritten as

$$\begin{cases} q_{11} = (\mathbf{I}z^n + \mathbf{A}_1z^{n-1} + \mathbf{A}_2z^{n-2} + \cdots + \mathbf{A}_n)^{-1}z^{n-1} \\ \quad = (\mathbf{I} + \mathbf{A}_1z^{-1} + \mathbf{A}_2z^{-2} + \cdots + \mathbf{A}_nz^{-n})^{-1}z^{-1} \\ \quad = \mathbf{A}^{-1}(z)z^{-1}, \\ q_{12} = q_{11}z^{-1}, \\ q_{13} = q_{11}z^{-2}, \\ \vdots \\ q_{1n} = q_{11}z^{-n+1}. \end{cases}$$

Using the properties of the shift operator, the first equation of (10) gives

$$z\mathbf{x}_0(t) = \mathbf{A}_0\mathbf{x}_0(t) + \mathbf{B}_0\mathbf{u}(t),$$

or

$$\mathbf{x}_0(t) = (z\mathbf{I} - \mathbf{A}_0)^{-1}\mathbf{B}_0\mathbf{u}(t).$$

Substituting $\mathbf{x}_0(t)$ into the second equation of (10) gives

$$\mathbf{y}(t) = \mathbf{C}_0(z\mathbf{I} - \mathbf{A}_0)^{-1}\mathbf{B}_0\mathbf{u}(t).$$

Thus we have the transfer matrix from the input $\mathbf{u}(t)$ to the output $\mathbf{y}(t)$:

$$\begin{aligned} \mathbf{G}(z) &= \mathbf{C}_0(z\mathbf{I} - \mathbf{A}_0)^{-1}\mathbf{B}_0 \\ &= [\mathbf{I}, \mathbf{0}, \mathbf{0}, \dots, \mathbf{0}] \begin{bmatrix} z\mathbf{I} + \mathbf{A}_1 & -\mathbf{I} & \mathbf{0} & \cdots & \mathbf{0} \\ \mathbf{A}_2 & z\mathbf{I} & -\mathbf{I} & \ddots & \vdots \\ \vdots & \mathbf{0} & \ddots & \ddots & \mathbf{0} \\ \mathbf{A}_{n-1} & \vdots & \ddots & z\mathbf{I} & -\mathbf{I} \\ \mathbf{A}_n & \mathbf{0} & \cdots & \mathbf{0} & z\mathbf{I} \end{bmatrix}^{-1} \begin{bmatrix} \mathbf{B}_1 \\ \mathbf{B}_2 \\ \vdots \\ \mathbf{B}_n \end{bmatrix} \\ &= [\mathbf{I}, \mathbf{0}, \mathbf{0}, \dots, \mathbf{0}] \begin{bmatrix} q_{11} & q_{12} & q_{13} & \cdots & q_{1n} \\ * & * & * & \cdots & * \\ \vdots & \vdots & \vdots & & \vdots \\ * & * & * & \cdots & * \end{bmatrix} \begin{bmatrix} \mathbf{B}_1 \\ \mathbf{B}_2 \\ \vdots \\ \mathbf{B}_n \end{bmatrix} \\ &= q_{11}\mathbf{B}_1 + q_{12}\mathbf{B}_2 + q_{13}\mathbf{B}_3 + \cdots + q_{1n}\mathbf{B}_n \\ &= q_{11}\mathbf{B}_1 + q_{11}\mathbf{B}_2z^{-1} + q_{11}\mathbf{B}_3z^{-2} + \cdots + q_{11}\mathbf{B}_nz^{-n+1} \\ &= q_{11}(\mathbf{B}_1 + \mathbf{B}_2z^{-1} + \mathbf{B}_3z^{-2} + \cdots + \mathbf{B}_nz^{-n+1}) \\ &= \mathbf{A}^{-1}(z)z^{-1}(\mathbf{B}_1 + \mathbf{B}_2z^{-1} + \mathbf{B}_3z^{-2} + \cdots + \mathbf{B}_nz^{-n+1}) \\ &= \mathbf{A}^{-1}(z)\mathbf{B}(z). \end{aligned}$$

Hence, we have the input–output representation:

$$\mathbf{A}(z)\mathbf{y}(t) = \mathbf{B}(z)\mathbf{u}(t).$$

This is equivalent to the difference equation in (1).

4. The observability canonical form from the difference equation

This section derives the observability canonical form from (1). Define new state variables,

$$\mathbf{x}_1(t) := \mathbf{y}(t), \quad (11)$$

$$\mathbf{x}_2(t) := \mathbf{y}(t+1) - \beta_1 \mathbf{u}(t), \quad (12)$$

$$\mathbf{x}_3(t) := \mathbf{y}(t+2) - \beta_1 \mathbf{u}(t+1) - \beta_2 \mathbf{u}(t), \quad (13)$$

\vdots

$$\mathbf{x}_n(t) := \mathbf{y}(t+n-1) - \beta_1 \mathbf{u}(t+n-2) - \beta_2 \mathbf{u}(t+n-3) - \cdots - \beta_{n-1} \mathbf{u}(t), \quad (14)$$

$$\mathbf{x}_n(t+1) := \mathbf{y}(t+n) - \beta_1 \mathbf{u}(t+n-1) - \beta_2 \mathbf{u}(t+n-2) - \cdots - \beta_{n-1} \mathbf{u}(t+1). \quad (15)$$

Subtracting $\beta_n \mathbf{u}(t)$ from both sides of (15) gives

$$\mathbf{x}_n(t+1) - \beta_n \mathbf{u}(t) = \mathbf{y}(t+n) - \beta_1 \mathbf{u}(t+n-1) - \beta_2 \mathbf{u}(t+n-2) - \cdots - \beta_{n-1} \mathbf{u}(t+1) - \beta_n \mathbf{u}(t). \quad (16)$$

Pre-multiplying (11)–(14) by the matrices $\mathbf{A}_n, \mathbf{A}_{n-1}, \mathbf{A}_{n-2}$ and \mathbf{A}_1 , respectively, gives

$$\begin{cases} \mathbf{A}_n \mathbf{x}_1(t) = \mathbf{A}_n \mathbf{y}(t), \\ \mathbf{A}_{n-1} \mathbf{x}_2(t) = \mathbf{A}_{n-1} \mathbf{y}(t+1) - \mathbf{A}_{n-1} \beta_1 \mathbf{u}(t), \\ \mathbf{A}_{n-2} \mathbf{x}_3(t) = \mathbf{A}_{n-2} \mathbf{y}(t+2) - \mathbf{A}_{n-2} \beta_1 \mathbf{u}(t+1) - \mathbf{A}_{n-2} \beta_2 \mathbf{u}(t), \\ \vdots \\ \mathbf{A}_1 \mathbf{x}_n(t) = \mathbf{A}_1 \mathbf{y}(t+n-1) - \mathbf{A}_1 \beta_1 \mathbf{u}(t+n-2) - \mathbf{A}_1 \beta_2 \mathbf{u}(t+n-3) - \cdots - \mathbf{A}_1 \beta_{n-1} \mathbf{u}(t). \end{cases} \quad (17)$$

Combining (16) and (17) gives

$$\begin{cases} \mathbf{A}_n \mathbf{y}(t) = \mathbf{A}_n \mathbf{x}_1(t), \\ \mathbf{A}_{n-1} \mathbf{y}(t+1) = \mathbf{A}_{n-1} \mathbf{x}_2(t) + \mathbf{A}_{n-1} \beta_1 \mathbf{u}(t), \\ \mathbf{A}_{n-2} \mathbf{y}(t+2) = \mathbf{A}_{n-2} \mathbf{x}_3(t) + \mathbf{A}_{n-2} \beta_1 \mathbf{u}(t+1) + \mathbf{A}_{n-2} \beta_2 \mathbf{u}(t), \\ \vdots \\ \mathbf{A}_1 \mathbf{y}(t+n-1) = \mathbf{A}_1 \mathbf{x}_n(t) + \mathbf{A}_1 \beta_1 \mathbf{u}(t+n-2) + \mathbf{A}_1 \beta_2 \mathbf{u}(t+n-3) + \cdots + \mathbf{A}_1 \beta_{n-1} \mathbf{u}(t), \\ \mathbf{y}(t+n) = \mathbf{x}_n(t+1) - \beta_n \mathbf{u}(t) + \beta_1 \mathbf{u}(t+n-1) + \beta_2 \mathbf{u}(t+n-2) + \cdots + \beta_n \mathbf{u}(t). \end{cases}$$

Adding the above equations gives

$$\begin{aligned} & \mathbf{y}(t+n) + \mathbf{A}_1 \mathbf{y}(t+n-1) + \mathbf{A}_2 \mathbf{y}(t+n-2) + \cdots + \mathbf{A}_n \mathbf{y}(t) \\ &= \mathbf{x}_n(t+1) - \beta_n \mathbf{u}(t) + \mathbf{A}_1 \mathbf{x}_n(t) + \mathbf{A}_2 \mathbf{x}_{n-1}(t) + \cdots + \mathbf{A}_n \mathbf{x}_1(t) \\ & \quad + \beta_1 \mathbf{u}(t+n-1) + (\mathbf{A}_1 \beta_1 + \beta_2) \mathbf{u}(t+n-2) + (\mathbf{A}_2 \beta_1 + \mathbf{A}_1 \beta_2 + \beta_3) \mathbf{u}(t+n-3) \\ & \quad + \cdots + (\mathbf{A}_{n-1} \beta_1 + \mathbf{A}_{n-2} \beta_2 + \mathbf{A}_{n-3} \beta_3 + \cdots + \beta_n) \mathbf{u}(t). \end{aligned}$$

Let

$$\begin{cases} \mathbf{B}_1 := \beta_1, \\ \mathbf{B}_2 := \mathbf{A}_1 \beta_1 + \beta_2, \\ \mathbf{B}_3 := \mathbf{A}_2 \beta_1 + \mathbf{A}_1 \beta_2 + \beta_3, \\ \vdots \\ \mathbf{B}_n := \mathbf{A}_{n-1} \beta_1 + \mathbf{A}_{n-2} \beta_2 + \mathbf{A}_{n-3} \beta_3 + \cdots + \beta_n, \end{cases}$$

which is equivalent to

$$\begin{bmatrix} \beta_1 \\ \beta_2 \\ \vdots \\ \beta_n \end{bmatrix} = \begin{bmatrix} \mathbf{I} & & & \\ \mathbf{A}_1 & \mathbf{I} & & \\ \mathbf{A}_2 & \mathbf{A}_1 & \mathbf{I} & \\ \vdots & & \ddots & \ddots \\ \mathbf{A}_{n-1} & \mathbf{A}_{n-2} & \cdots & \mathbf{A}_1 & \mathbf{I} \end{bmatrix}^{-1} \begin{bmatrix} \mathbf{B}_1 \\ \mathbf{B}_2 \\ \vdots \\ \mathbf{B}_n \end{bmatrix}.$$

Replacing t in (1) with $t+n$ yields

$$\begin{aligned} & \mathbf{y}(t+n) + \mathbf{A}_1 \mathbf{y}(t+n-1) + \mathbf{A}_2 \mathbf{y}(t+n-2) + \cdots + \mathbf{A}_n \mathbf{y}(t) \\ &= \mathbf{B}_1 \mathbf{u}(t+n-1) + \mathbf{B}_2 \mathbf{u}(t+n-2) + \mathbf{B}_3 \mathbf{u}(t+n-3) + \cdots + \mathbf{B}_n \mathbf{u}(t). \end{aligned}$$

Adding Eq. (16) to (17) gives

$$\begin{aligned}
 & \mathbf{x}_n(t+1) - \beta_n \mathbf{u}(t) + \mathbf{A}_1 \mathbf{x}_n(t) + \mathbf{A}_2 \mathbf{x}_{n-1}(t) + \cdots + \mathbf{A}_n \mathbf{x}_1(t) \\
 &= \mathbf{y}(t+n) - \beta_1 \mathbf{u}(t+n-1) - \beta_2 \mathbf{u}(t+n-2) - \cdots - \beta_n \mathbf{u}(t) \\
 &\quad + \mathbf{A}_1 [\mathbf{y}(t+n-1) - \beta_1 \mathbf{u}(t+n-2) - \beta_2 \mathbf{u}(t+n-3) - \cdots - \beta_{n-1} \mathbf{u}(t)] \\
 &\quad + \mathbf{A}_2 [\mathbf{y}(t+n-2) - \beta_1 \mathbf{u}(t+n-3) - \beta_2 \mathbf{u}(t+n-4) - \cdots - \beta_{n-2} \mathbf{u}(t)] + \cdots + \mathbf{A}_n \mathbf{y}(t) \\
 &= \mathbf{y}(t+n) + \mathbf{A}_1 \mathbf{y}(t+n-1) + \mathbf{A}_2 \mathbf{y}(t+n-2) + \cdots + \mathbf{A}_n \mathbf{y}(t) \\
 &\quad - \beta_1 \mathbf{u}(t+n-1) - (\mathbf{A}_1 \beta_1 + \beta_2) \mathbf{u}(t+n-2) - (\mathbf{A}_2 \beta_1 + \mathbf{A}_1 \beta_2 + \beta_3) \mathbf{u}(t+n-3) \\
 &\quad - \cdots - (\mathbf{A}_{n-1} \beta_1 + \mathbf{A}_{n-2} \beta_2 + \mathbf{A}_{n-3} \beta_3 + \cdots + \beta_n) \mathbf{u}(t) \\
 &= \mathbf{y}(t+n) + \mathbf{A}_1 \mathbf{y}(t+n-1) + \mathbf{A}_2 \mathbf{y}(t+n-2) + \cdots + \mathbf{A}_n \mathbf{y}(t) \\
 &\quad - \mathbf{B}_1 \mathbf{u}(t+n-1) - \mathbf{B}_2 \mathbf{u}(t+n-2) - \mathbf{B}_3 \mathbf{u}(t+n-3) - \cdots - \mathbf{B}_n \mathbf{u}(t) \\
 &= \mathbf{0},
 \end{aligned}$$

or

$$\mathbf{x}_n(t+1) - \beta_n \mathbf{u}(t) = -\mathbf{A}_n \mathbf{x}_1(t) - \mathbf{A}_{n-1} \mathbf{x}_2(t) - \mathbf{A}_{n-2} \mathbf{x}_3(t) - \cdots - \mathbf{A}_1 \mathbf{x}_n(t). \quad (18)$$

Replacing t in (11)–(14) with $t+1$ and combining (18) gives

$$\begin{cases}
 \mathbf{x}_1(t+1) = \mathbf{y}(t+1) \\
 \quad = \mathbf{x}_2(t) + \beta_1 \mathbf{u}(t), \\
 \mathbf{x}_2(t+1) = \mathbf{y}(t+2) - \beta_1 \mathbf{u}(t+1) \\
 \quad = \mathbf{x}_3(t) + \beta_2 \mathbf{u}(t), \\
 \vdots \\
 \mathbf{x}_{n-1}(t+1) = \mathbf{y}(t+n-1) - \beta_1 \mathbf{u}(t+n-2) - \beta_2 \mathbf{u}(t+n-3) - \cdots - \beta_{n-2} \mathbf{u}(t+1) \\
 \quad = \mathbf{x}_n(t) + \beta_{n-1} \mathbf{u}(t), \\
 \mathbf{x}_n(t+1) = -\mathbf{A}_n \mathbf{x}_1(t) - \mathbf{A}_{n-1} \mathbf{x}_2(t) - \mathbf{A}_{n-2} \mathbf{x}_3(t) - \cdots - \mathbf{A}_1 \mathbf{x}_n(t) + \beta_n \mathbf{u}(t).
 \end{cases}$$

Combining (11) and the above equations gives the following state space model,

$$\begin{cases}
 \begin{bmatrix} \mathbf{x}_1(t+1) \\ \mathbf{x}_2(t+1) \\ \vdots \\ \mathbf{x}_n(t+1) \end{bmatrix} = \begin{bmatrix} \mathbf{0} & \mathbf{I} & \mathbf{0} & \cdots & \mathbf{0} \\ \mathbf{0} & \mathbf{0} & \mathbf{I} & \ddots & \vdots \\ \vdots & \vdots & \ddots & \ddots & \mathbf{0} \\ \mathbf{0} & \mathbf{0} & \cdots & \mathbf{0} & \mathbf{I} \\ -\mathbf{A}_n & -\mathbf{A}_{n-1} & -\mathbf{A}_{n-2} & \cdots & -\mathbf{A}_1 \end{bmatrix} \begin{bmatrix} \mathbf{x}_1(t) \\ \mathbf{x}_2(t) \\ \vdots \\ \mathbf{x}_n(t) \end{bmatrix} + \begin{bmatrix} \beta_1 \\ \beta_2 \\ \vdots \\ \beta_n \end{bmatrix} \mathbf{u}(t), \\
 \mathbf{y}(t) = [\mathbf{I}, \mathbf{0}, \mathbf{0}, \dots, \mathbf{0}] \begin{bmatrix} \mathbf{x}_1(t) \\ \mathbf{x}_2(t) \\ \vdots \\ \mathbf{x}_n(t) \end{bmatrix},
 \end{cases} \quad (19)$$

where

$$\begin{bmatrix} \beta_1 \\ \beta_2 \\ \vdots \\ \beta_n \end{bmatrix} = \begin{bmatrix} \mathbf{I} & & & & \\ \mathbf{A}_1 & \mathbf{I} & & & \\ \mathbf{A}_2 & \mathbf{A}_1 & \mathbf{I} & & \\ \vdots & \vdots & \ddots & \ddots & \\ \mathbf{A}_{n-1} & \mathbf{A}_{n-2} & \cdots & \mathbf{A}_1 & \mathbf{I} \end{bmatrix}^{-1} \begin{bmatrix} \mathbf{B}_1 \\ \mathbf{B}_2 \\ \vdots \\ \mathbf{B}_n \end{bmatrix}.$$

Or

$$\begin{cases}
 \mathbf{x}_{\text{ob}}(t+1) = \mathbf{A}_{\text{ob}} \mathbf{x}_{\text{ob}}(t) + \mathbf{B}_{\text{ob}} \mathbf{u}(t), \\
 \mathbf{y}(t) = \mathbf{C}_{\text{ob}} \mathbf{x}_{\text{ob}}(t),
 \end{cases}$$

$$\mathbf{A}_{\text{ob}} := \begin{bmatrix} \mathbf{0} & \mathbf{I} & \mathbf{0} & \cdots & \mathbf{0} \\ \mathbf{0} & \mathbf{0} & \mathbf{I} & \ddots & \vdots \\ \vdots & \vdots & \ddots & \ddots & \mathbf{0} \\ \mathbf{0} & \mathbf{0} & \cdots & \mathbf{0} & \mathbf{I} \\ -\mathbf{A}_n & -\mathbf{A}_{n-1} & -\mathbf{A}_{n-2} & \cdots & -\mathbf{A}_1 \end{bmatrix} \in \mathbb{R}^{(mn) \times (mn)}, \quad \mathbf{B}_{\text{ob}} := \begin{bmatrix} \beta_1 \\ \beta_2 \\ \vdots \\ \beta_n \end{bmatrix} \in \mathbb{R}^{(mn) \times r},$$

$$\mathbf{C}_{\text{ob}} := [\mathbf{I}, \mathbf{0}, \mathbf{0}, \dots, \mathbf{0}] \in \mathbb{R}^{m \times (mn)}, \quad \mathbf{A}_i \in \mathbb{R}^{m \times m}, \quad \beta_i \in \mathbb{R}^{m \times r},$$

where $\mathbf{x}_{\text{ob}}(t) := [\mathbf{x}_1(t), \mathbf{x}_2(t), \dots, \mathbf{x}_n(t)]^T \in \mathbb{R}^{mn}$ is the state vector, \mathbf{A}_{ob} is the bottom-row companion matrix, \mathbf{B}_{ob} is an arbitrary matrix and \mathbf{C}_{ob} is a matrix whose the first $m \times m$ block is an identity matrix and the rest is a zero matrix.

This state space model with \mathbf{A}_{ob} and \mathbf{C}_{ob} special structures is called the observability canonical form.

5. The observability canonical form to the difference equation

In this section, from the observability canonical form in (19), we derive its input–output representation, which does not involve the state vector $\mathbf{x}(t)$.

Let

$$\mathbf{P} := z\mathbf{I} - \mathbf{A}_{\text{ob}} = \begin{bmatrix} z\mathbf{I} & -\mathbf{I} & \mathbf{0} & \cdots & \mathbf{0} \\ \mathbf{0} & z\mathbf{I} & -\mathbf{I} & \ddots & \vdots \\ \vdots & \ddots & \ddots & \ddots & \mathbf{0} \\ \mathbf{0} & \cdots & \mathbf{0} & z\mathbf{I} & -\mathbf{I} \\ \mathbf{A}_n & \mathbf{A}_{n-1} & \cdots & \mathbf{A}_2 & z\mathbf{I} + \mathbf{A}_1 \end{bmatrix}, \quad \mathbf{P}^{-1} := \begin{bmatrix} \mathbf{p}_{11} & \mathbf{p}_{12} & \mathbf{p}_{13} & \cdots & \mathbf{p}_{1n} \\ * & * & * & \cdots & * \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ * & * & * & \cdots & * \end{bmatrix}.$$

According to $\mathbf{P}^{-1}\mathbf{P} = \mathbf{I}$, we have

$$\begin{bmatrix} \mathbf{p}_{11} & \mathbf{p}_{12} & \mathbf{p}_{13} & \cdots & \mathbf{p}_{1n} \\ * & * & * & \cdots & * \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ * & * & * & \cdots & * \end{bmatrix} \begin{bmatrix} z\mathbf{I} & -\mathbf{I} & \mathbf{0} & \cdots & \mathbf{0} \\ \mathbf{0} & z\mathbf{I} & -\mathbf{I} & \ddots & \vdots \\ \vdots & \ddots & \ddots & \ddots & \mathbf{0} \\ \mathbf{0} & \cdots & \mathbf{0} & z\mathbf{I} & -\mathbf{I} \\ \mathbf{A}_n & \mathbf{A}_{n-1} & \cdots & \mathbf{A}_2 & z\mathbf{I} + \mathbf{A}_1 \end{bmatrix} = \begin{bmatrix} \mathbf{I} & \mathbf{0} & \cdots & \mathbf{0} \\ \mathbf{0} & \mathbf{I} & \cdots & \mathbf{0} \\ \vdots & \vdots & \ddots & \vdots \\ \mathbf{0} & \mathbf{0} & \cdots & \mathbf{I} \end{bmatrix}.$$

Expanding this matrix equation, its first row gives the following equations:

$$\begin{cases} \mathbf{p}_{11}z + \mathbf{p}_{1n}\mathbf{A}_n = \mathbf{I}, \\ -\mathbf{p}_{11} + \mathbf{p}_{12}z + \mathbf{p}_{1n}\mathbf{A}_{n-1} = \mathbf{0}, \\ -\mathbf{p}_{12} + \mathbf{p}_{13}z + \mathbf{p}_{1n}\mathbf{A}_{n-2} = \mathbf{0}, \\ \vdots \\ -\mathbf{p}_{1n-1} + \mathbf{p}_{1n}(z\mathbf{I} + \mathbf{A}_1) = \mathbf{0}, \end{cases}$$

which can be equivalently rewritten as

$$\begin{aligned} \mathbf{p}_{1n} &= (\mathbf{I}z^n + \mathbf{A}_1z^{n-1} + \mathbf{A}_2z^{n-2} + \cdots + \mathbf{A}_n)^{-1}, \\ \mathbf{p}_{1n-1} &= \mathbf{p}_{1n}(\mathbf{I}z + \mathbf{A}_1), \\ &\vdots \\ \mathbf{p}_{13} &= \mathbf{p}_{1n}(\mathbf{I}z^{n-3} + \mathbf{A}_1z^{n-4} + \mathbf{A}_2z^{n-5} + \cdots + \mathbf{A}_{n-3}), \\ \mathbf{p}_{12} &= \mathbf{p}_{1n}(\mathbf{I}z^{n-2} + \mathbf{A}_1z^{n-3} + \mathbf{A}_2z^{n-4} + \cdots + \mathbf{A}_{n-2}), \\ \mathbf{p}_{11} &= \mathbf{p}_{1n}(\mathbf{I}z^{n-1} + \mathbf{A}_1z^{n-2} + \mathbf{A}_2z^{n-3} + \cdots + \mathbf{A}_{n-1}). \end{aligned}$$

The transfer matrix from the input $\mathbf{u}(t)$ to the output $\mathbf{y}(t)$ is

$$\begin{aligned} \mathbf{G}(z) &= \mathbf{C}_{\text{ob}}(z\mathbf{I} - \mathbf{A}_{\text{ob}})^{-1}\mathbf{B}_{\text{ob}} \\ &= [\mathbf{I}, \mathbf{0}, \mathbf{0}, \dots, \mathbf{0}] \begin{bmatrix} z\mathbf{I} & -\mathbf{I} & \mathbf{0} & \cdots & \mathbf{0} \\ \mathbf{0} & z\mathbf{I} & -\mathbf{I} & \ddots & \vdots \\ \vdots & \ddots & \ddots & \ddots & \mathbf{0} \\ \mathbf{0} & \cdots & \mathbf{0} & z\mathbf{I} & -\mathbf{I} \\ \mathbf{A}_n & \mathbf{A}_{n-1} & \cdots & \mathbf{A}_2 & z\mathbf{I} + \mathbf{A}_1 \end{bmatrix}^{-1} \begin{bmatrix} \beta_1 \\ \beta_2 \\ \vdots \\ \beta_n \end{bmatrix} \\ &= [\mathbf{I}, \mathbf{0}, \mathbf{0}, \dots, \mathbf{0}] \begin{bmatrix} \mathbf{p}_{11} & \mathbf{p}_{12} & \mathbf{p}_{13} & \cdots & \mathbf{p}_{1n} \\ * & * & * & \cdots & * \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ * & * & * & \cdots & * \end{bmatrix} \begin{bmatrix} \beta_1 \\ \beta_2 \\ \vdots \\ \beta_n \end{bmatrix} \\ &= \mathbf{p}_{11}\beta_1 + \mathbf{p}_{12}\beta_2 + \mathbf{p}_{13}\beta_3 + \cdots + \mathbf{p}_{1n}\beta_n \\ &= \mathbf{p}_{1n}[(\mathbf{I}z^{n-1} + \mathbf{A}_1z^{n-2} + \mathbf{A}_2z^{n-3} + \cdots + \mathbf{A}_{n-1})\beta_1 + (\mathbf{I}z^{n-2} + \mathbf{A}_1z^{n-3} + \mathbf{A}_2z^{n-4} + \cdots + \mathbf{A}_{n-2})\beta_2 \\ &\quad + (\mathbf{I}z^{n-3} + \mathbf{A}_1z^{n-4} + \mathbf{A}_2z^{n-5} + \cdots + \mathbf{A}_{n-3})\beta_3 + \cdots + \beta_n] \end{aligned}$$

$$\begin{aligned}
&= (\mathbf{I}z^n + \mathbf{A}_1 z^{n-1} + \mathbf{A}_2 z^{n-2} + \cdots + \mathbf{A}_n)^{-1} [(\mathbf{I}z^{n-1} + \mathbf{A}_1 z^{n-2} + \mathbf{A}_2 z^{n-3} + \cdots + \mathbf{A}_{n-1})\boldsymbol{\beta}_1 \\
&\quad + (\mathbf{I}z^{n-2} + \mathbf{A}_1 z^{n-3} + \mathbf{A}_2 z^{n-4} + \cdots + \mathbf{A}_{n-2})\boldsymbol{\beta}_2 + (\mathbf{I}z^{n-3} + \mathbf{A}_1 z^{n-4} + \mathbf{A}_2 z^{n-5} + \cdots + \mathbf{A}_{n-3})\boldsymbol{\beta}_3 + \cdots + \boldsymbol{\beta}_n] \\
&= \mathbf{A}^{-1}(z)\mathbf{B}(z).
\end{aligned}$$

Hence, we have the input–output representation:

$$\mathbf{A}(z)\mathbf{y}(t) = \mathbf{B}(z)\mathbf{u}(t),$$

where

$$\begin{bmatrix} \boldsymbol{\beta}_1 \\ \boldsymbol{\beta}_2 \\ \vdots \\ \boldsymbol{\beta}_n \end{bmatrix} = \begin{bmatrix} \mathbf{I} & & & & \\ \mathbf{A}_1 & \mathbf{I} & & & \\ \mathbf{A}_2 & \mathbf{A}_1 & \mathbf{I} & & \\ \vdots & \vdots & \ddots & \ddots & \\ \mathbf{A}_{n-1} & \mathbf{A}_{n-2} & \cdots & \mathbf{A}_1 & \mathbf{I} \end{bmatrix}^{-1} \begin{bmatrix} \mathbf{B}_1 \\ \mathbf{B}_2 \\ \vdots \\ \mathbf{B}_n \end{bmatrix}.$$

This is equivalent to the difference equation in (1).

6. Examples

Example 1. Consider the following difference equation:

$$\mathbf{y}(t) + \begin{bmatrix} 1 & 2 \\ 3 & 4 \end{bmatrix} \mathbf{y}(t-1) + \begin{bmatrix} 5 & 6 \\ 7 & 8 \end{bmatrix} \mathbf{y}(t-2) = \begin{bmatrix} 9 & 10 \\ 11 & 12 \end{bmatrix} \mathbf{u}(t-1) + \begin{bmatrix} 13 & 14 \\ 15 & 16 \end{bmatrix} \mathbf{u}(t-2).$$

Using the formula in (9), its observer canonical form is given by

$$\begin{cases} \mathbf{x}(t+1) = \begin{bmatrix} -1 & -2 & 1 & 0 \\ -3 & -4 & 0 & 1 \\ -5 & -6 & 0 & 0 \\ -7 & -8 & 0 & 0 \end{bmatrix} \mathbf{x}(t) + \begin{bmatrix} 9 & 10 \\ 11 & 12 \\ 13 & 14 \\ 15 & 16 \end{bmatrix} \mathbf{u}(t), \\ \mathbf{y}(t) = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix} \mathbf{x}(t). \end{cases}$$

Compute the parameters:

$$\begin{bmatrix} \boldsymbol{\beta}_1 \\ \boldsymbol{\beta}_2 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 1 & 2 & 1 & 0 \\ 3 & 4 & 0 & 1 \end{bmatrix}^{-1} \begin{bmatrix} 9 & 10 \\ 11 & 12 \\ 13 & 14 \\ 15 & 16 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ -1 & -2 & 1 & 0 \\ -3 & -4 & 0 & 1 \end{bmatrix} \begin{bmatrix} 9 & 10 \\ 11 & 12 \\ 13 & 14 \\ 15 & 16 \end{bmatrix} = \begin{bmatrix} 9 & 10 \\ 11 & 12 \\ -18 & -20 \\ -56 & -62 \end{bmatrix}.$$

Using the formula in (19), its observability canonical form is given by

$$\begin{cases} \mathbf{x}(t+1) = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -5 & -6 & -1 & -2 \\ -7 & -8 & -3 & -4 \end{bmatrix} \mathbf{x}(t) + \begin{bmatrix} \boldsymbol{\beta}_1 \\ \boldsymbol{\beta}_2 \end{bmatrix} \mathbf{u}(t), \\ \mathbf{y}(t) = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix} \mathbf{x}(t), \end{cases}$$

Example 2. The following observer canonical form:

$$\begin{cases} \mathbf{x}(t+1) = \begin{bmatrix} -1 & -2 & 1 & 0 \\ -3 & -4 & 0 & 1 \\ -5 & -6 & 0 & 0 \\ -7 & -8 & 0 & 0 \end{bmatrix} \mathbf{x}(t) + \begin{bmatrix} 9 & 10 \\ 11 & 12 \\ 13 & 14 \\ 15 & 16 \end{bmatrix} \mathbf{u}(t), \\ \mathbf{y}(t) = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix} \mathbf{x}(t), \end{cases}$$

and observability canonical form:

$$\begin{cases} \mathbf{x}(t+1) = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -5 & -6 & -1 & -2 \\ -7 & -8 & -3 & -4 \end{bmatrix} \mathbf{x}(t) + \begin{bmatrix} 9 & 10 \\ 11 & 12 \\ -18 & -20 \\ -56 & -62 \end{bmatrix} \mathbf{u}(t), \\ \mathbf{y}(t) = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix} \mathbf{x}(t) \end{cases}$$

corresponding the difference equation:

$$\mathbf{y}(t) + \begin{bmatrix} 1 & 2 \\ 3 & 4 \end{bmatrix} \mathbf{y}(t-1) + \begin{bmatrix} 5 & 6 \\ 7 & 8 \end{bmatrix} \mathbf{y}(t-2) = \begin{bmatrix} 9 & 10 \\ 11 & 12 \end{bmatrix} \mathbf{u}(t-1) + \begin{bmatrix} 13 & 14 \\ 15 & 16 \end{bmatrix} \mathbf{u}(t-2).$$

7. Conclusions

This paper derives the observer canonical state space model and observability canonical state space model from difference equations and vice versa, by setting different state variables. Two examples are provided to demonstrate the proposed methods.

References

- [1] J.S.H. Tsai, C.T. Wang, C.C. Kuang, S.M. Guo, L.S. Shieh, C.W. Chen, A NARMAX model-based state-space self-tuning control for nonlinear stochastic hybrid systems, *Applied Mathematical Modelling* 34 (10) (2010) 3030–3054.
- [2] B. Yan, G.W. Yan, A steady-state lattice Boltzmann model for incompressible flows, *Computers and Mathematics with Applications* 61 (5) (2011) 1348–1354.
- [3] F. Ding, T. Chen, Least squares based self-tuning control of dual-rate systems, *International Journal of Adaptive Control and Signal Processing* 18 (8) (2004) 697–714.
- [4] F. Ding, T. Chen, Z. Iwai, Adaptive digital control of Hammerstein nonlinear systems with limited output sampling, *SIAM Journal on Control and Optimization* 45 (6) (2007) 2257–2276.
- [5] F. Ding, T. Chen, A gradient based adaptive control algorithm for dual-rate systems, *Asian Journal of Control* 8 (4) (2006) 314–323.
- [6] A. Gelişken, C. Çinar, A.S. Kurbanli, On the asymptotic behavior and periodic nature of a difference equation with maximum, *Computers and Mathematics with Applications* 59 (2) (2010) 898–902.
- [7] G.P. Wei, J.H. Shen, Boundedness and asymptotic behavior results for nonlinear difference equations with positive and negative coefficients, *Computers and Mathematics with Applications* 60 (8) (2010) 2469–2475.
- [8] A. Ashyralyev, Y. Sözen, Well-posedness of parabolic differential and difference equations, *Computers and Mathematics with Applications* 60 (3) (2010) 792–802.
- [9] A. Kaushik, K.K. Sharma, M. Sharma, A parameter uniform difference scheme for parabolic partial differential equation with a retarded argument, *Applied Mathematical Modelling* 34 (12) (2010) 4232–4242.
- [10] B. Li, Q.K. Song, Asymptotic behaviors of non-autonomous impulsive difference equation with delays, *Applied Mathematical Modelling* 35 (7) (2011) 3423–3433.
- [11] D. Kumar, A parameter-uniform numerical method for time-dependent singularly perturbed differential-difference equations, *Applied Mathematical Modelling* 35 (6) (2011) 2805–2819.
- [12] M.K. Kadalbajoo, D. Kumar, A computational method for singularly perturbed nonlinear differential-difference equations with small shift, *Applied Mathematical Modelling* 34 (9) (2010) 2584–2596.
- [13] A.S. Kurbanli, C. Çinar, İ. Yalçinkaya, On the behavior of positive solutions of the system of rational difference equations, *Mathematical and Computer Modelling* 53 (5–6) (2011) 1261–1267.
- [14] B.D. Iričanin, Global stability of some classes of higher-order nonlinear difference equations, *Applied Mathematics and Computation* 216 (4) (2010) 1325–1328.
- [15] T.B. Schön, A. Wills, B. Ninness, System identification of nonlinear state-space models, *Automatica* 47 (1) (2011) 39–49.
- [16] M. Rachid, B. Maamar, D. Said, Multivariable fractional system approximation with initial conditions using integral state space representation, *Computers and Mathematics with Applications* 59 (5) (2010) 1842–1851.
- [17] F. Ding, T. Chen, Hierarchical identification of lifted state-space models for general dual-rate systems, *IEEE Transactions on Circuits and Systems–I: Regular Papers* 52 (6) (2005) 1179–1187.
- [18] F. Ding, L. Qiu, T. Chen, Reconstruction of continuous-time systems from their non-uniformly sampled discrete-time systems, *Automatica* 45 (2) (2009) 324–332.
- [19] L. Xie, Y.J. Liu, H.Z. Yang, F. Ding, Modelling and identification for non-uniformly periodically sampled-data systems, *IET Control Theory and Applications* 4 (5) (2010) 784–794.
- [20] L. Xie, H.Z. Yang, F. Ding, Recursive least squares parameter estimation for non-uniformly sampled systems based on the data filtering, *Mathematical and Computer Modelling* 54 (1–2) (2011) 315–324.
- [21] Y. Shi, H. Fang, Kalman filter based identification for systems with randomly missing measurements in a network environment, *International Journal of Control* 83 (3) (2010) 538–551.
- [22] H. Fang, J. Wu, Y. Shi, Genetic adaptive state estimation with missing input/output data, *Proceedings of the Institution of Mechanical Engineers, Part I: Journal of Systems and Control Engineering* 224 (5) (2010) 611–617.
- [23] F. Ding, T. Chen, Hierarchical least squares identification methods for multivariable systems, *IEEE Transactions on Automatic Control* 50 (3) (2005) 397–402.
- [24] F. Ding, T. Chen, Hierarchical gradient-based identification of multivariable discrete-time systems, *Automatica* 41 (2) (2005) 315–325.
- [25] F. Ding, H.B. Chen, M. Li, Multi-innovation least squares identification methods based on the auxiliary model for MISO systems, *Applied Mathematics and Computation* 187 (2) (2007) 658–668.
- [26] F. Ding, Y.S. Xiao, A finite-data-window least squares algorithm with a forgetting factor for dynamical modeling, *Applied Mathematics and Computation* 186 (1) (2007) 184–192.
- [27] F. Ding, G. Liu, X.P. Liu, Partially coupled stochastic gradient identification methods for non-uniformly sampled systems, *IEEE Transactions on Automatic Control* 55 (8) (2010) 1976–1981.

- [28] D.Q. Wang, F. Ding, Extended stochastic gradient identification algorithms for Hammerstein–Wiener ARMAX systems, *Computers & Mathematics with Applications* 56 (12) (2008) 3157–3164.
- [29] Z.N. Zhang, F. Ding, X.G. Liu, Hierarchical gradient based iterative parameter estimation algorithm for multivariable output error moving average systems, *Computers and Mathematics with Applications* 61 (3) (2011) 672–682.
- [30] Y.J. Liu, Y.S. Xiao, X.L. Zhao, Multi-innovation stochastic gradient algorithm for multiple-input single-output systems using the auxiliary model, *Applied Mathematics and Computation* 215 (4) (2009) 1477–1483.
- [31] L.L. Han, F. Ding, Multi-innovation stochastic gradient algorithms for multi-input multi-output systems, *Digital Signal Processing* 19 (4) (2009) 545–554.
- [32] Y.J. Liu, J. Sheng, R.F. Ding, Convergence of stochastic estimation gradient algorithm for multivariable ARX-like systems, *Computers and Mathematics with Applications* 59 (8) (2010) 2615–2627.
- [33] Y. Zhang, G.M. Cui, Bias compensation methods for stochastic systems with colored noise, *Applied Mathematical Modelling* 35 (4) (2011) 1709–1716.
- [34] Y. Zhang, Unbiased identification of a class of multi-input single-output systems with correlated disturbances using bias compensation methods, *Mathematical and Computer Modelling* 53 (9–10) (2011) 1810–1819.
- [35] J. Ding, Y.J. Liu, F. Ding, Iterative solutions to matrix equations of form $AiXBi = Fi$, *Computers and Mathematics with Applications* 59 (11) (2010) 3500–3507.
- [36] M. Dehghan, M. Hajarian, Analysis of an iterative algorithm to solve the generalized coupled Sylvester matrix equations, *Applied Mathematical Modelling* 35 (7) (2011) 3285–3300.
- [37] M. Dehghan, M. Hajarian, Computing matrix functions using mixed interpolation methods, *Mathematical and Computer Modelling* 52 (5–6) (2010) 826–836.
- [38] F. Ding, T. Chen, Parameter estimation of dual-rate stochastic systems by using an output error method, *IEEE Transactions on Automatic Control* 50 (9) (2005) 1436–1441.
- [39] F. Ding, Y. Shi, T. Chen, Performance analysis of estimation algorithms of non-stationary ARMA processes, *IEEE Transactions on Signal Processing* 54 (3) (2006) 1041–1053.
- [40] F. Ding, T. Chen, Performance bounds of the forgetting factor least squares algorithm for time-varying systems with finite measurement data, *IEEE Transactions on Circuits and Systems–I: Regular Papers* 52 (3) (2005) 555–566.
- [41] F. Ding, T. Chen, Identification of Hammerstein nonlinear ARMAX systems, *Automatica* 41 (9) (2005) 1479–1489.
- [42] F. Ding, Y. Shi, T. Chen, Gradient-based identification methods for Hammerstein nonlinear ARMAX models, *Nonlinear Dynamics* 45 (1–2) (2006) 31–43.
- [43] F. Ding, Y. Shi, T. Chen, Auxiliary model based least-squares identification methods for Hammerstein output-error systems, *Systems & Control Letters* 56 (5) (2007) 373–380.
- [44] J. Ding, F. Ding, X.P. Liu, G. Liu, Hierarchical least squares identification for linear SISO systems with dual-rate sampled-data, *IEEE Transactions on Automatic Control* 56 (11) (2011) 2677–2683.
- [45] F. Ding, P.X. Liu, H.Z. Yang, Parameter identification and intersample output estimation for dual-rate systems, *IEEE Transactions on Systems, Man, and Cybernetics, Part A: Systems and Humans* 38 (4) (2008) 966–975.
- [46] J. Ding, F. Ding, Bias compensation based parameter estimation for output error moving average systems, *International Journal of Adaptive Control and Signal Processing* 25 (12) (2011) 1100–1111.
- [47] F. Ding, Y.J. Liu, B. Bao, Gradient based and least squares based iterative estimation algorithms for multi-input multi-output systems, *Proceedings of the Institution of Mechanical Engineers, Part I: Journal of Systems and Control Engineering* 226 (1) (2012) 43–55.
- [48] F. Ding, T. Chen, L. Qiu, Bias compensation based recursive least squares identification algorithm for MISO systems, *IEEE Transactions on Circuits and Systems–II: Express Briefs* 53 (5) (2006) 349–353.
- [49] F. Ding, T. Chen, Performance analysis of multi-innovation gradient type identification methods, *Automatica* 43 (1) (2007) 1–14.
- [50] F. Ding, P.X. Liu, G. Liu, Auxiliary model based multi-innovation extended stochastic gradient parameter estimation with colored measurement noises, *Signal Processing* 89 (10) (2009) 1883–1890.
- [51] J.B. Zhang, F. Ding, Y. Shi, Self-tuning control based on multi-innovation stochastic gradient parameter estimation, *Systems & Control Letters* 58 (1) (2009) 69–75.
- [52] F. Ding, Several multi-innovation identification methods, *Digital Signal Processing* 20 (4) (2010) 1027–1039.
- [53] D.Q. Wang, F. Ding, Performance analysis of the auxiliary models based multi-innovation stochastic gradient estimation algorithm for output error systems, *Digital Signal Processing* 20 (3) (2010) 750–762.
- [54] F. Ding, G. Liu, X.P. Liu, Parameter estimation with scarce measurements, *Automatica* 47 (8) (2011) 1646–1655.
- [55] F. Ding, P.X. Liu, G. Liu, Multi-innovation least squares identification for linear and pseudo-linear regression models, *IEEE Transactions on Systems, Man, and Cybernetics, Part B: Cybernetics* 40 (3) (2010) 767–778.
- [56] Y.J. Liu, L. Yu, F. Ding, Multi-innovation extended stochastic gradient algorithm and its performance analysis, *Circuits, Systems and Signal Processing* 29 (4) (2010) 649–667.
- [57] J.H. Li, F. Ding, Maximum likelihood stochastic gradient estimation for Hammerstein systems with colored noise based on the key term separation technique, *Computers & Mathematics with Applications* 62 (11) (2011) 4170–4177.
- [58] W. Wang, J.H. Li, R.F. Ding, Maximum likelihood parameter estimation algorithm for controlled autoregressive autoregressive models, *International Journal of Computer Mathematics* 88 (16) (2011) 3458–3467.
- [59] J.H. Li, F. Ding, G.W. Yang, Maximum likelihood least squares identification method for input nonlinear finite impulse response moving average systems, *Mathematical and Computer Modelling* 55 (3–4) (2012) 442–450.
- [60] Y.J. Liu, D.Q. Wang, F. Ding, Least-squares based iterative algorithms for identifying Box–Jenkins models with finite measurement data, *Digital Signal Processing* 20 (5) (2010) 1458–1467.
- [61] D.Q. Wang, G.W. Yang, R.F. Ding, Gradient-based iterative parameter estimation for Box–Jenkins systems, *Computers & Mathematics with Applications* 60 (5) (2010) 1200–1208.
- [62] H.Q. Han, L. Xie, F. Ding, X.G. Liu, Hierarchical least squares based iterative identification for multivariable systems with moving average noises, *Mathematical and Computer Modelling* 51 (9–10) (2010) 1213–1220.
- [63] F. Ding, P.X. Liu, G. Liu, Gradient based and least-squares based iterative identification methods for OE and OEMA systems, *Digital Signal Processing* 20 (3) (2010) 664–677.
- [64] F. Ding, P.X. Liu, G. Liu, Identification methods for Hammerstein nonlinear systems, *Digital Signal Processing* 21 (2) (2011) 215–238.
- [65] D.Q. Wang, F. Ding, Least squares based and gradient based iterative identification for Wiener nonlinear systems, *Signal Processing* 91 (5) (2011) 1182–1189.
- [66] M. Dehghan, M. Hajarian, Finite iterative algorithms for the reflexive and anti-reflexive solutions of the matrix equation $A_1X_1B_1 + A_2X_2B_2 = C$, *Mathematical and Computer Modelling* 49 (9–10) (2009) 1937–1959.
- [67] D.Q. Wang, Least squares-based recursive and iterative estimation for output error moving average (OEMA) systems using data filtering, *IET Control Theory and Applications* 5 (14) (2011) 1648–1657.
- [68] Y.J. Liu, L. Xie, F. Ding, An auxiliary model based recursive least squares parameter estimation algorithm for non-uniformly sampled multirate systems, *Proceedings of the Institution of Mechanical Engineers, Part I: Journal of Systems and Control Engineering* 223 (4) (2009) 445–454.
- [69] F. Ding, T. Chen, Identification of dual-rate systems based on finite impulse response models, *International Journal of Adaptive Control and Signal Processing* 18 (7) (2004) 589–598.
- [70] F. Ding, J. Ding, Least squares parameter estimation with irregularly missing data, *International Journal of Adaptive Control and Signal Processing* 24 (7) (2010) 540–553.

- [71] Y. Shi, B. Yu, Robust mixed H_2/H_∞ control of networked control systems with random time delays in both forward and backward communication links, *Automatica* 47 (4) (2011) 754–760.
- [72] Y. Shi, F. Ding, T. Chen, 2-Norm based recursive design of transmultiplexers with designable filter length, *Circuits, Systems and Signal Processing* 25 (4) (2006) 447–462.
- [73] D.Q. Wang, F. Ding, Input–output data filtering based recursive least squares parameter estimation for CARARMA systems, *Digital Signal Processing* 20 (4) (2010) 991–999.
- [74] F. Ding, Transformations between some special matrices, *Computers & Mathematics with Applications* 59 (8) (2010) 2676–2695.
- [75] F. Ding, P.X. Liu, J. Ding, Iterative solutions of the generalized Sylvester matrix equations by using the hierarchical identification principle, *Applied Mathematics and Computation* 197 (1) (2008) 41–50.